



Louis Van Lommel & Raf Vleugels

WÜRTH ELEKTRONIK MORE THAN YOU EXPECT

WÜRTH ELEKTRONIK eiSos GROUP



PASSIVE
COMPONENTS



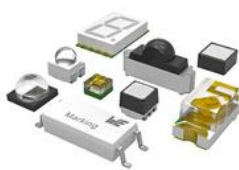
POWER MODULES
(MAGI³C-SERIES)



DIGITAL ISOLATORS



OPTOELECTRONICS



THERMAL
MANAGEMENT



ELECTROMECHANICAL
COMPONENTS



RADIO MODULES &
SENSORS



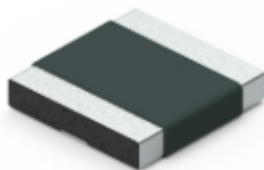
AUTOMOTIVE





Gap Filling Solutions

WE-TGF Thermal Gap Filler Pad | WE-TINS Thermally Conductive Insulator Pad |
WE-TTT Thermal Transfer Tape | WE-PCM Thermal Phase Changing Material



Hybrid Solutions

WE-TGFG Graphite Foam Gasket



Heat Spreading Solutions

WE-TGS Graphite Sheet

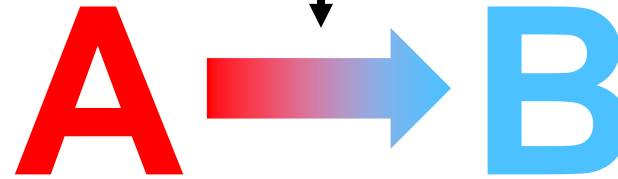


Heat Sinks **NEW**

WE-HTO Heatsink for TO220/247 |
WE-HTOI Heatsink for TO220/247 with Integrated Interface Material | [+ 3 more](#)



Thermal Management Solutions

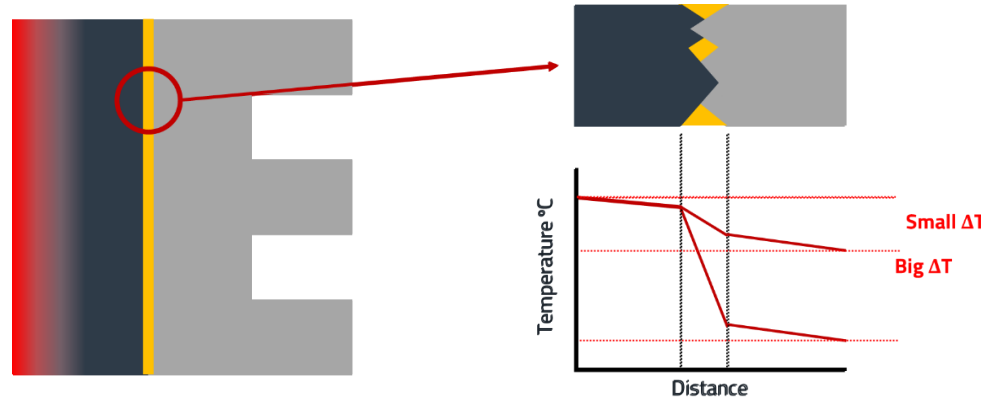


Heating Element

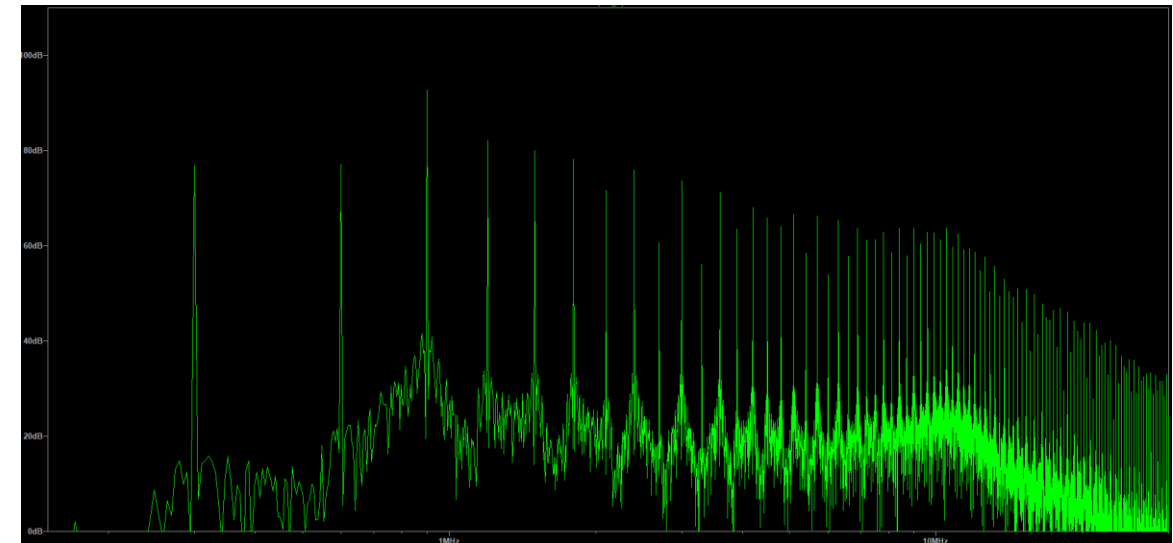
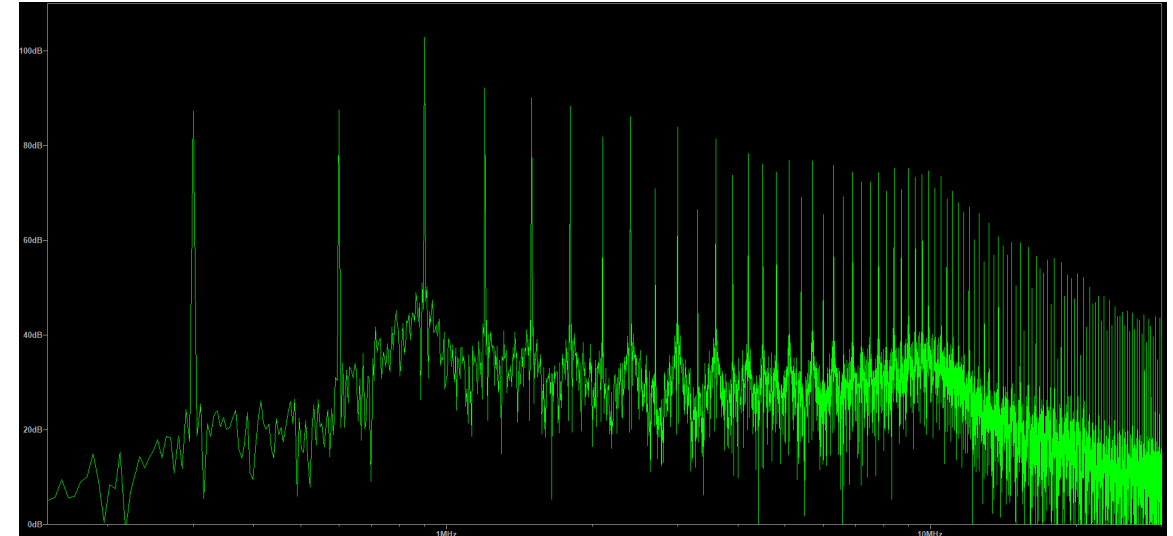
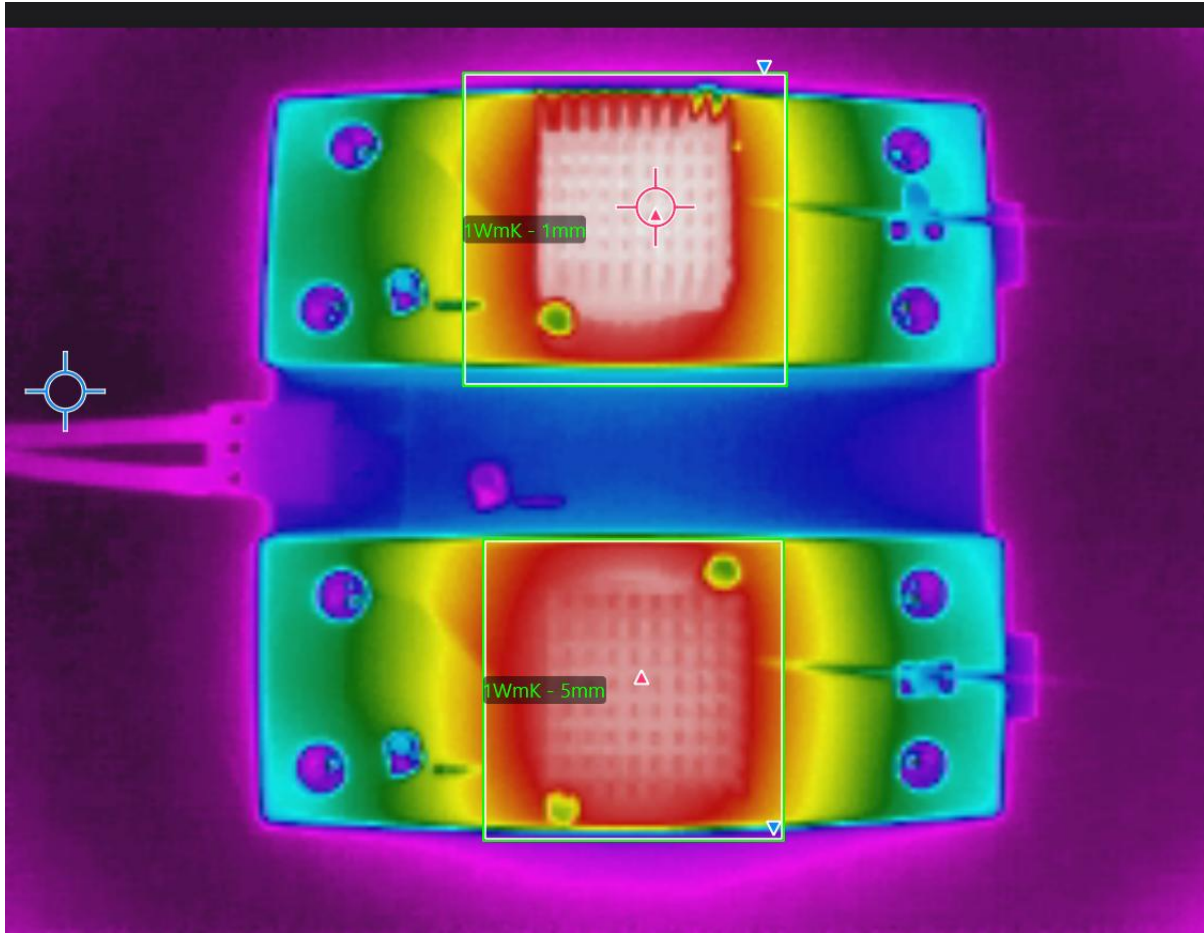
- One hot component
- Multiple hot components
- Back side of a PCB
- Processing units
- Transistors, IGBTs
- Memory ICs
- Battery cells
- ...

Cooling Assembly

- Heatsink
- Metal housing
- Cooling plate
- Radiator
- ...



Today's Demo:

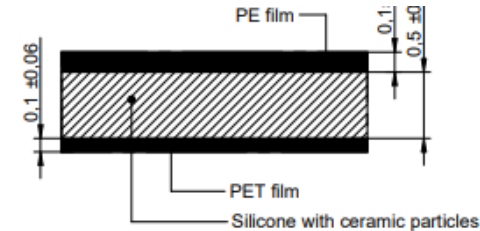


Thermal Resistance of TGF

$$R_{th-TIM} = \frac{l}{\lambda \cdot A} = \frac{0,002m}{1 \frac{W}{m \cdot K} \cdot (0,031 * 0,031)} = 2,08 \frac{K}{W}$$

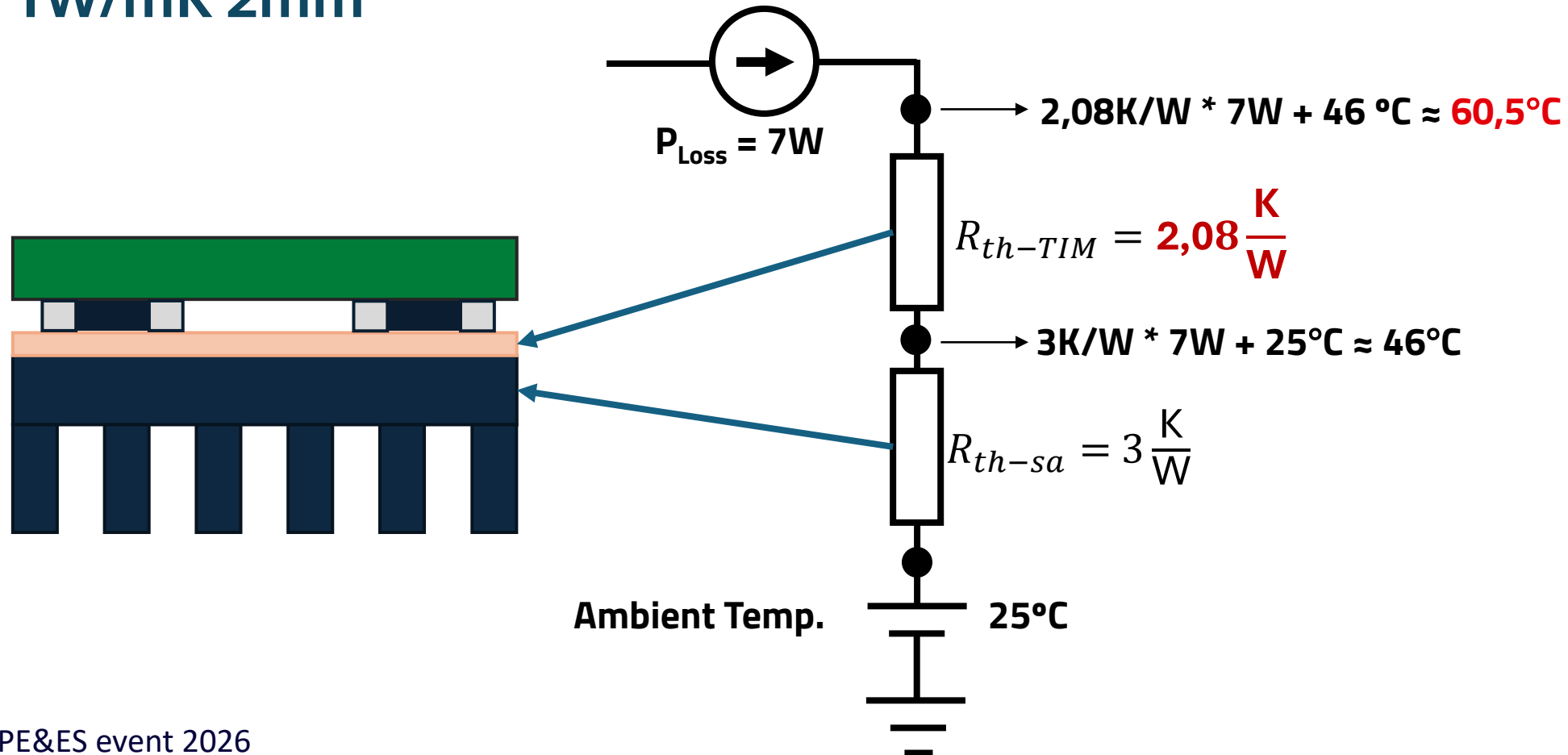
$$R_{th-TIM} = \frac{l}{\lambda \cdot A} = \frac{0,005m}{1 \frac{W}{m \cdot K} \cdot (0,031 * 0,031)} = 5,20 \frac{K}{W}$$

- l = Material thickness of TIM[m]
- A = Contact surface area of TIM[m²]
- λ = Thermal conductivity coefficient of TIM [$\frac{W}{m \cdot K}$]



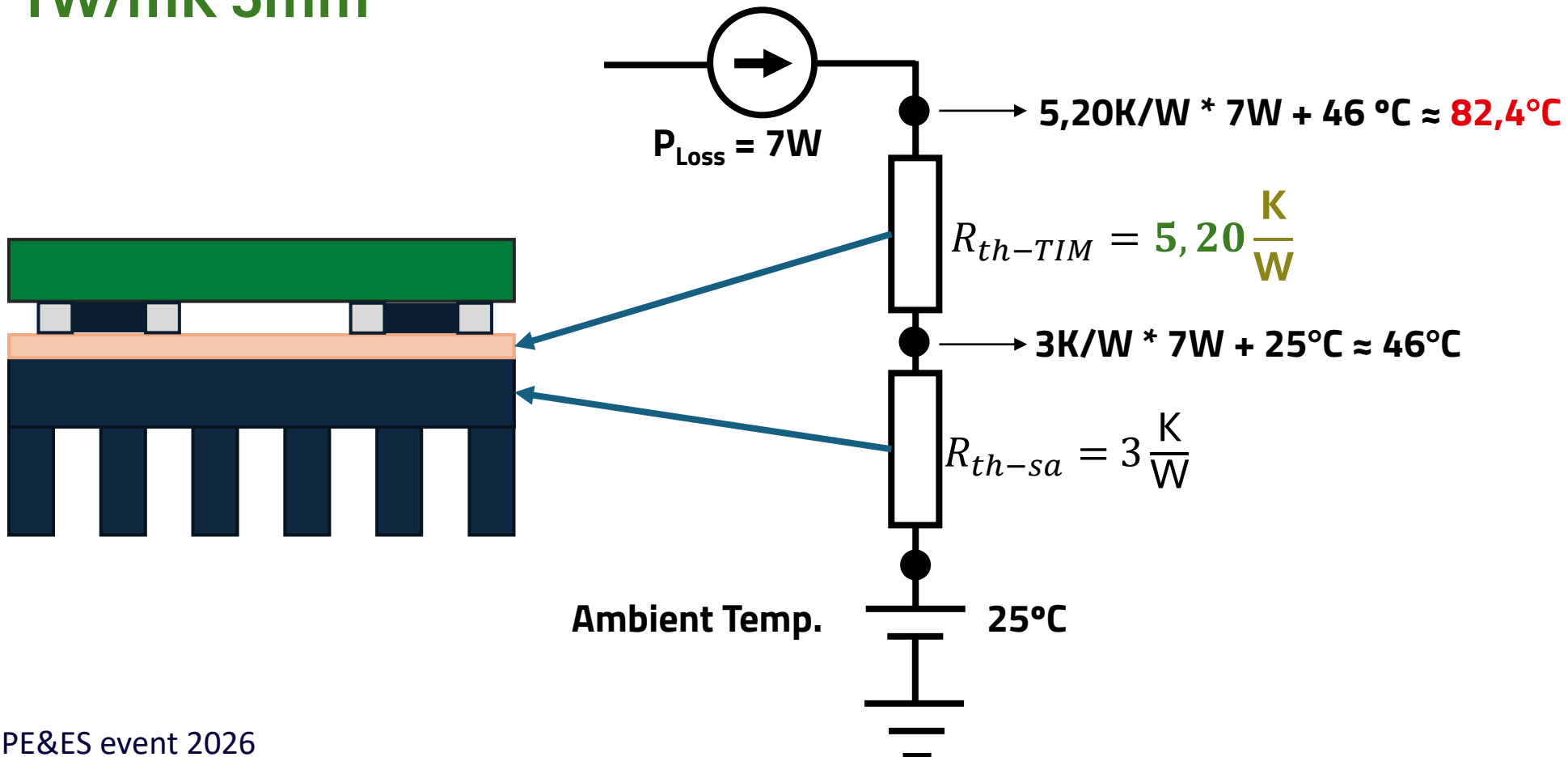
- **Case 2: passive heatsink with TGF**

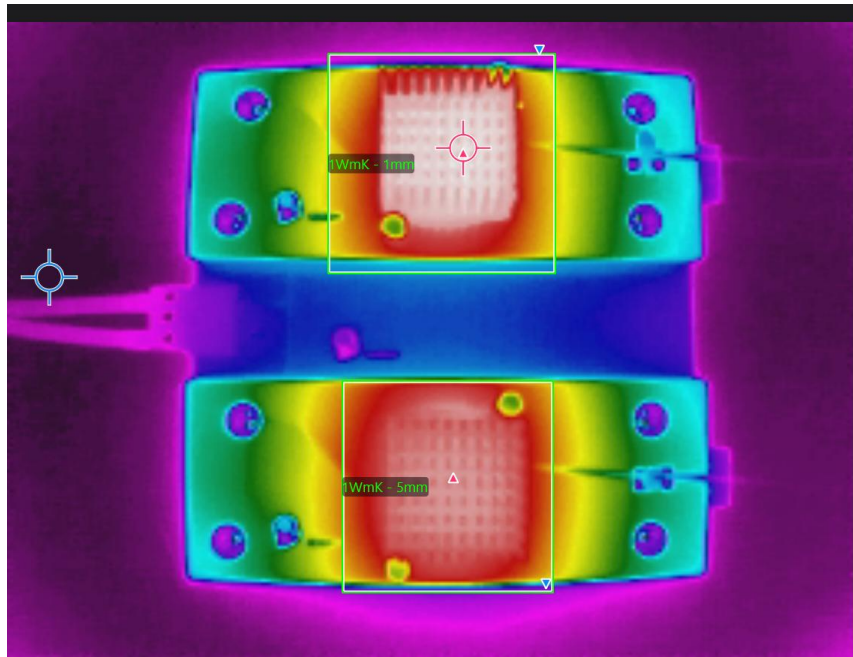
1W/mK 2mm



- **Case 2: passive heatsink with TGF**

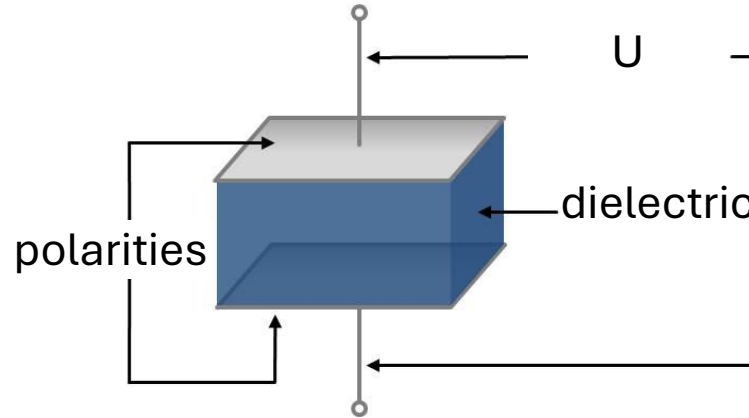
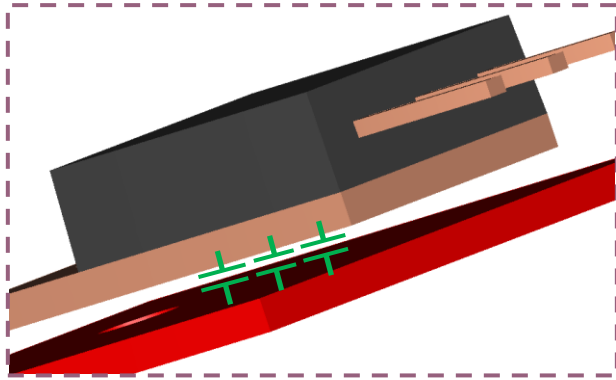
1W/mK 5mm





TMI	T°C Component		T°C Heatsink	
	Calculation	Measurement	Calculation	Measurement
1W/mK 2mm	60,5	66	46	62
1W/mK 5mm	82,4	84	46	60

EMC Aspects - Thermal interface is not only a thermal path, but also electrical coupling path.



$$C = \varepsilon * \frac{A}{d} = \varepsilon_0 * \varepsilon_r * \frac{A}{d}$$

C – Capacitance [F]

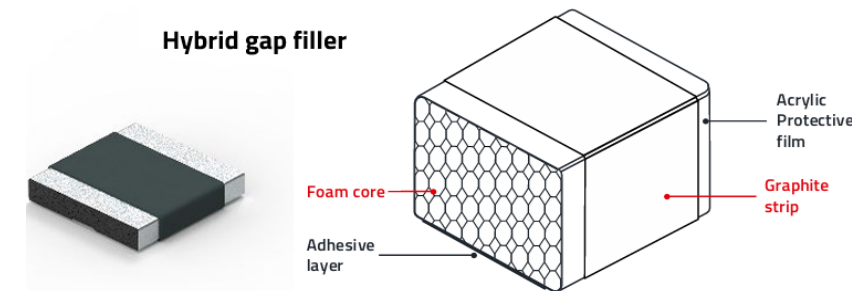
A – plate area [m²]

d – plate distance [m]

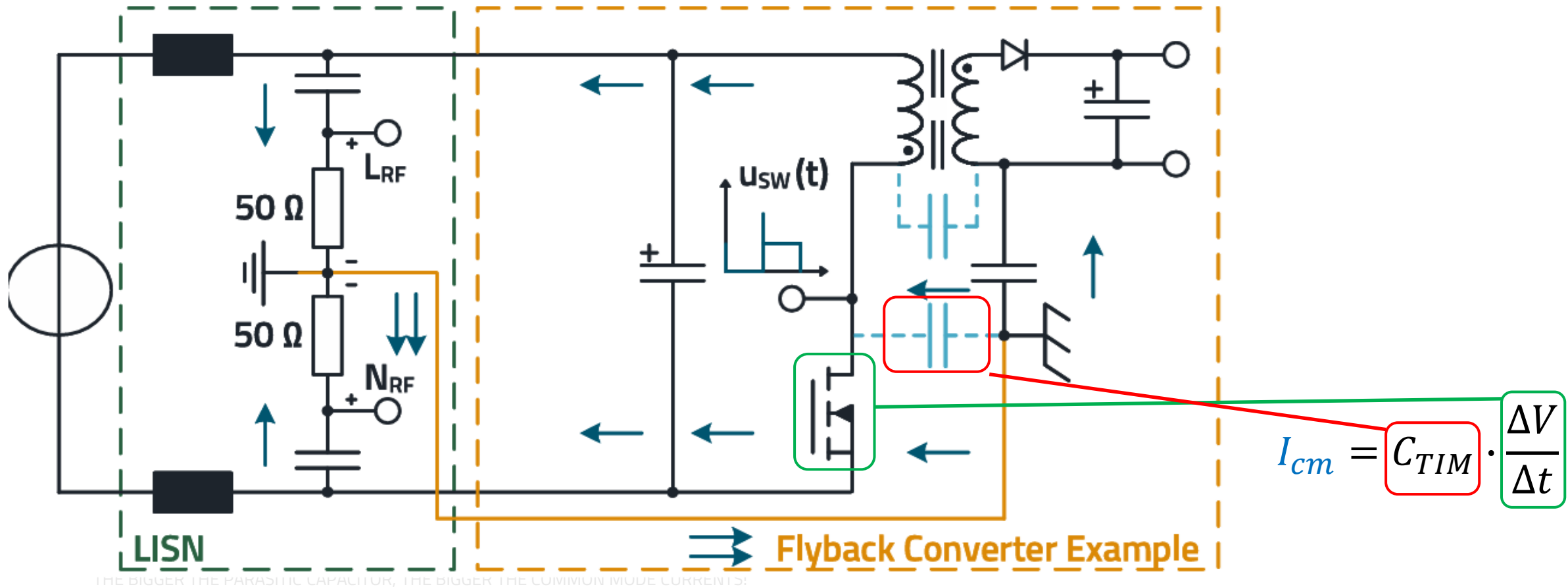
ε_r – relative permittivity TIM

ε₀ –absolute permittivity : $8,85 \times 10^{-12} \left[\frac{As}{Vm} \right]$

Silicone WE-TGF 5mm k = 3			Silicone WE-TGF 5mm k = 1			Silicone WE-TGF 2mm k = 1			Hybrid WE-TGFG 5mm		
Permittivity	13		Permittivity	7,8		Permittivity	7,8		Permittivity		
	8,85E-12			8,85E-12			8,85E-12			8,85E-12	
A	0,0004 m ²		A	0,0004 m ²		A	0,0004 m ²		A	0,0004 m ²	
d	0,004 m		d	0,004 m		d	0,0016 m		d	0,004 m	
Calculated C	1,15E-11	11,51 pF	Calculated C	6,90E-12	6,90 pF	Calculated C	1,73E-11	17,26 pF	Calculated C		pF
Measured C		11,4 pF	Measured C		7,5 pF	Measured C		13,5 pF	Measured C		81 pF



EMC Aspects - Parasitic capacitance creates common-Mode Current in a Flyback converter



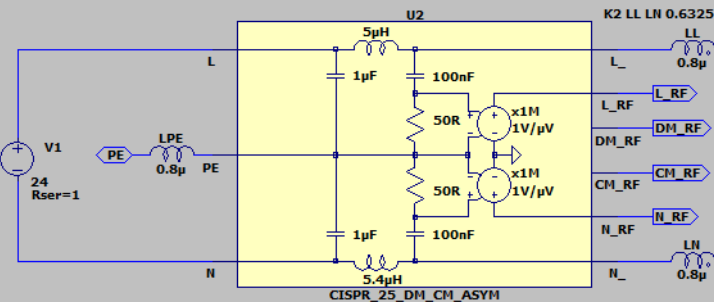
THE BIGGER THE PARASITIC CAPACITANCE, THE BIGGER THE COMMON MODE CURRENTS!

THERMAL COUPLING CAN UNINTENTIONALLY BECOME AN EMI COUPLING PATH

Simulation settings

```
.lib us1g.lib
.lib SMAJ.lib
.lib SMC303.lib
.model SW SW(Ron=1m Roff=1Meg Vt=0.5 Vh=-0.5)
.tran 0 1.193333m 1m 10n
.options abstol=0.01
.options numdgt=7
.options plotwinsize=0 fastaccess
.options noopiter gminsteps=0
.save V(DM_RF) V(CM_RF)
.ic V(IN)=24
```

Supply + Asym LISN! + Line



Note

1. Run simulation
2. View FFT
3. Select V(cm_rf), V(dm_rf) and V(l_rf) by holding the ctrl-key - OK
4. Plot Settings - Reload Plot Settings

LPE = Line inductance PE
 LL = Line inductance L/+
 LN = Line inductance N/-
 LLK = Primary-side leakage inductance
 CDE = Drain to Earth (PE) capacitor
 CPS = Primary to secondary winding capacitor (interwinding capacitor)
 RPS = primary to secondary resistor (find the DC operating point)

Line inductance measured with LCR45 (1mm=1nH)
 $K2 = \sqrt{(1-LSC/LC)} = \sqrt{(1-0.48\mu H/0.8\mu H)} = 0.6325$

Filter

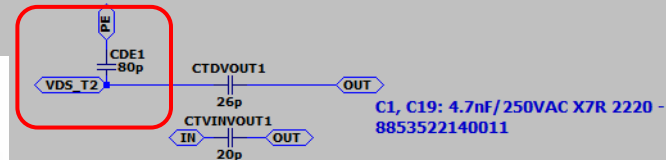
CMC: 2.2mH/2A - 74482222
 C2, C15: 4.7nF/250VAC X7R 2220 - 8853522140011
 C12: 4.7µF/50V X7R 1210 - 885012209048
 D6: TVSP 30V/3kW D0-214AB - 824551301



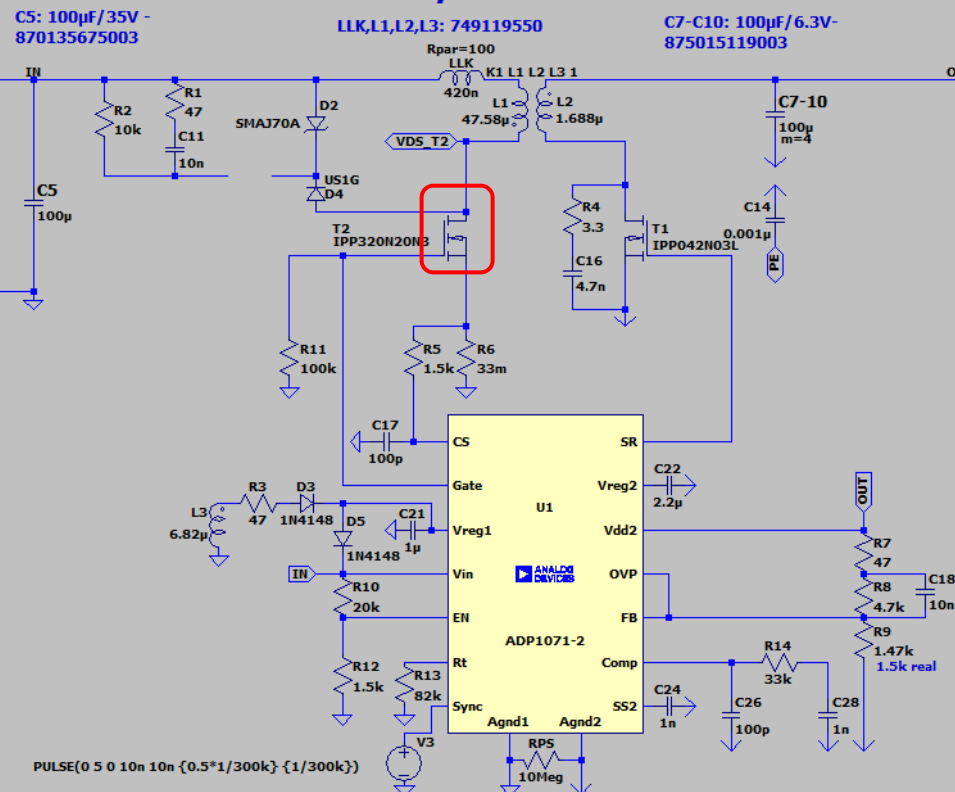
80pF



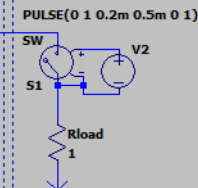
Common-Mode Noise Model!

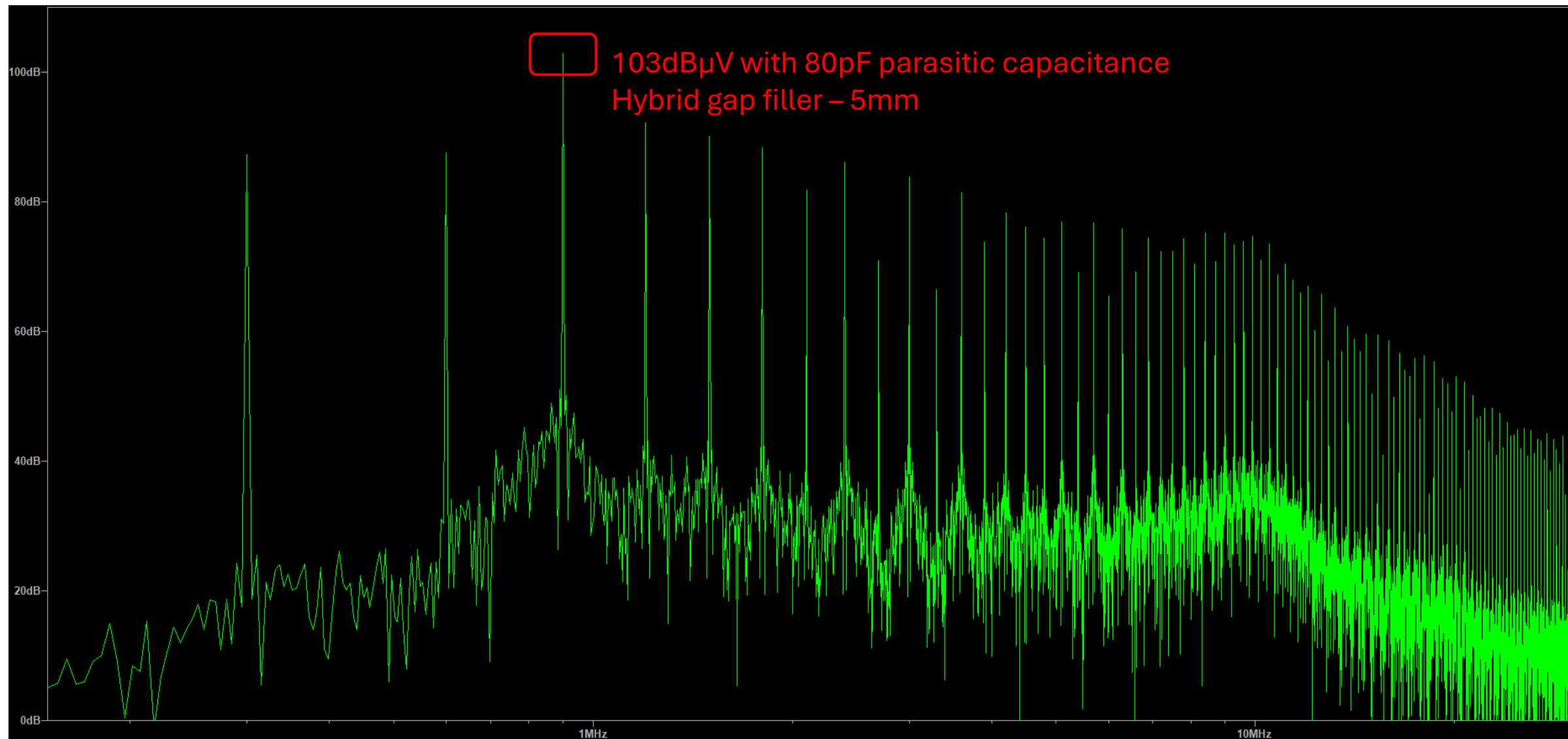


Flyback-Converter



Load



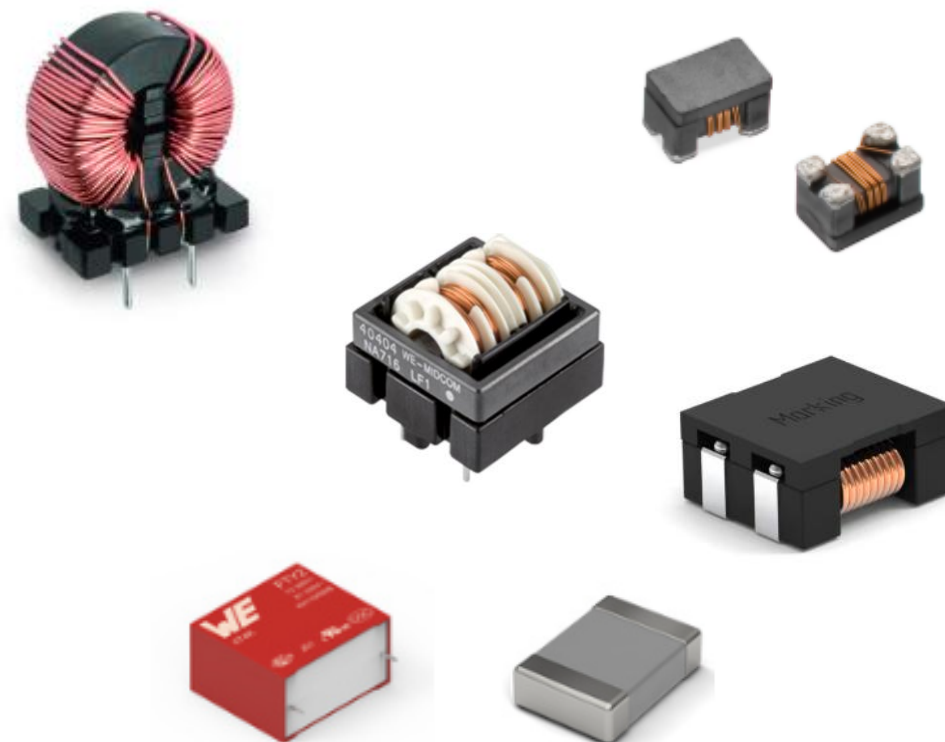


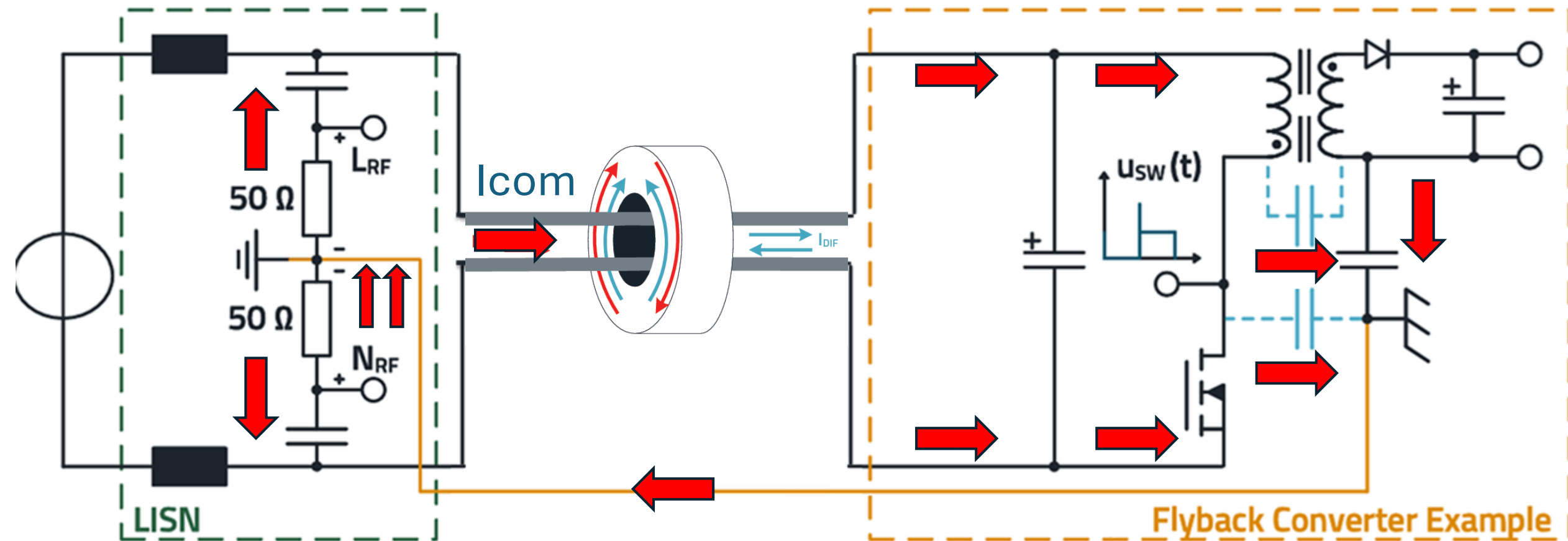
	Thermal Resistance	Capacitance	db μ V @ 900kHz
5mm Graphite over Foam	Low	80pF	103
5mm TIM – k = 3	Medium	11,5pF	94
5mm TIM – k = 1	High	7,5pF	93
2mm TIM – k = 1	Medium	13,5pF	94,5

System-Level / Mechanical Measures

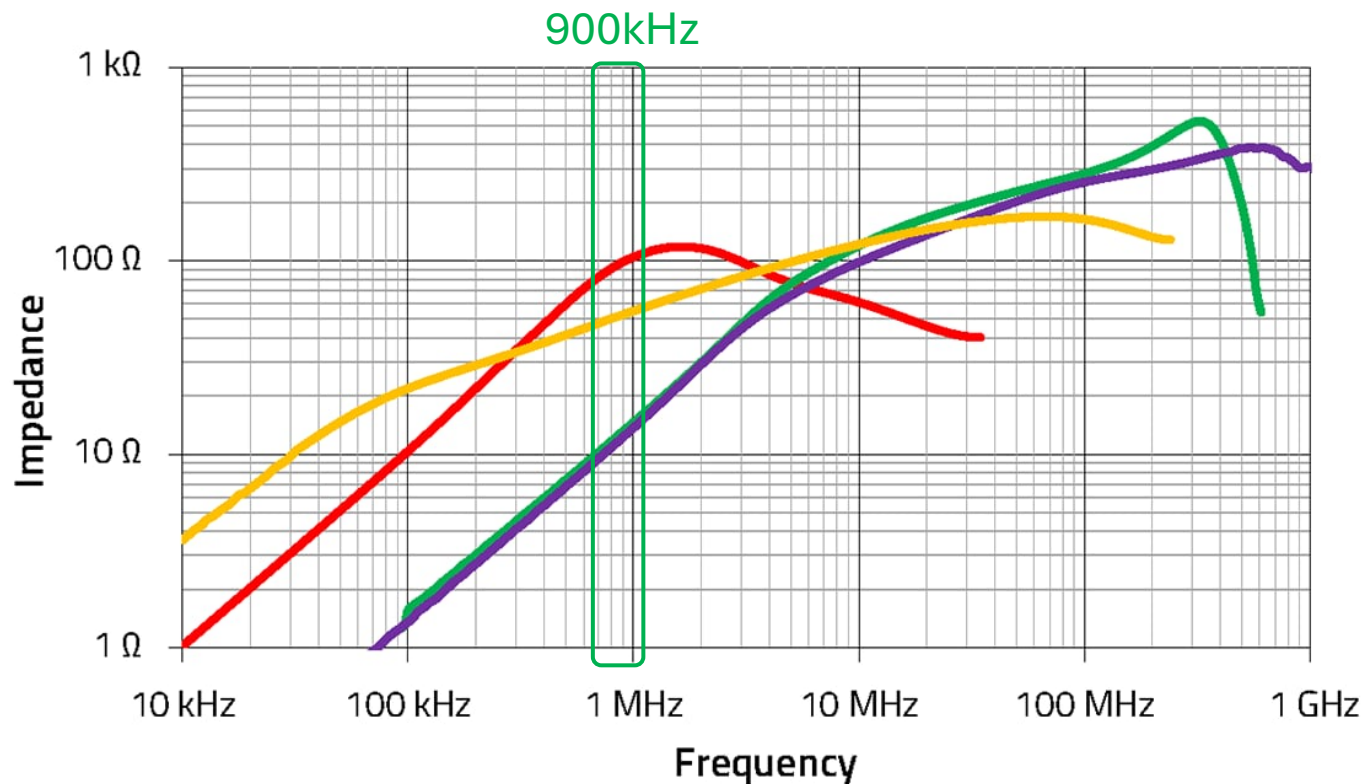


PCB-Level / Circuit Measures

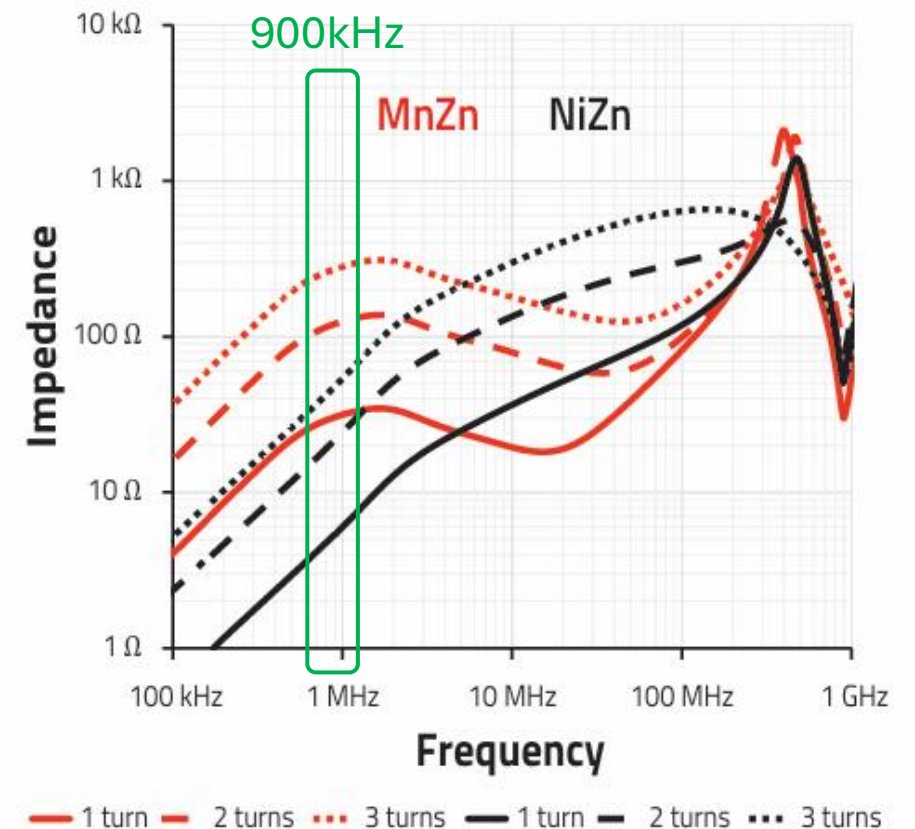




Common mode solutions - Cable ferrites



— MnZn — NiZn — MgZn — NC



— 1 turn — 2 turns ... 3 turns — 1 turn — 2 turns ... 3 turns



Design considerations - Thermal Performance versus EMI considerations

Thermal

- Thinner TIM = Better thermal performance
- High conductivity = Better thermal performance

$$R_{th} = \frac{l}{\lambda \cdot A} = \left[\frac{K}{W} \right]$$



$$\lambda = \text{Thermal conductivity} \left[\frac{W}{m \cdot K} \right]$$

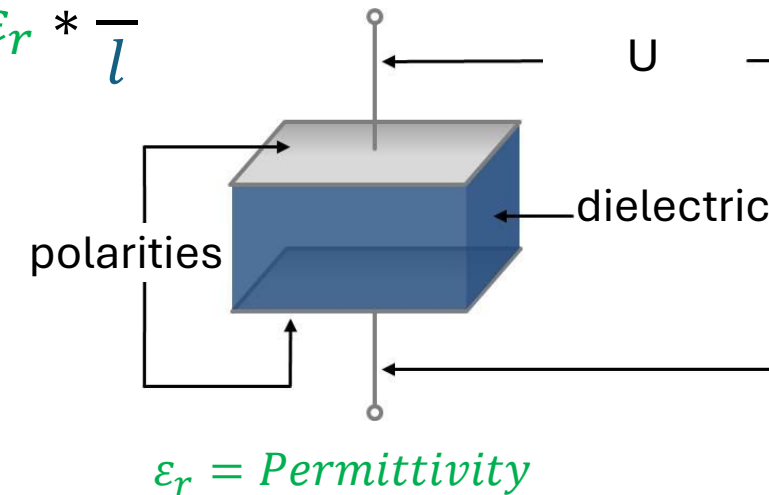
$$l = \text{Material thickness} [m]$$

$$A = \text{Cooling overall surface} [m^2]$$

EMI

- Thicker TIM = Better EMI performance
- Lower permittivity = Better EMI performance

$$C = \epsilon_0 * \epsilon_r * \frac{A}{l}$$



- Do not select a TIM on thermal conductivity alone.
- Check thickness, area and permittivity as EMC-relevant parameters.
- If redesign is difficult, use system-level mitigation such as cable ferrites.
- Come check out our demo!



